A Compact Printed End-Fire Antenna for Radio Frequency Identification (RFID) Handheld Reader

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Abstract - A new compact end-fire linear polarized printed antenna for a handheld radio frequency identification (RFID) reader is proposed for ultra high frequency (UHF) radio frequency identification (RFID) system in North America. The proposed planar end-fire antenna using two meandered dipole drivers, a folded reflector and a rectangular reflector are presented. The new antenna uses low-cost fabrication. The advantage of the end-fire antenna with meander dipole drivers compared to the conventional quasi-Yagi antenna is a reduction in the length of the driver, which allows closer space for RFID reader, the dimension of the antenna is $80 \times 59 \text{ mm}^2$. The antenna has maximum gain of 3.6 dB and VSWR better than 2 around the US RFID bands (902-928MHz). We describe the antenna structure and present the comparison of simulation results with experimental data. The proposed antenna is fabricated, and measured reflection coefficient, radiation patterns and gain are presented.

Index Terms – RFID Reader, end-fire antenna, Radiation pattern, Gain.

I. INTRODUCTION

In recent years, radio frequency identification (RFID) technology has been rapidly developed and applied to many service industries,

manufacturing companies, distribution logistics, goods flow systems and moving vehicle identification[1-3]. Many typical RFID tags have been studied[4-7]. For the applications involving item-level management[8][9], a RFID handheld reader plays an important role owing to its advantages of compactness, flexibility and maneuverability. By incorporating with a personal data assistant (PDA), a RFID handheld reader has the ability to provide a total solution for retail or library automation management. The growing demand for small compact wireless devices has increased the need for small antennas that can be integrated while providing acceptable overall performance[10][11]. Most of antennas do not have directional radiation. Some antennas have bidirectional radiation, but the its size is big for handheld reader [12]. In addition, usability of the reader unit in terms of reading directions and orientations of tags has to be taken into account. One of the features affecting the size, weight and ergonomist of handheld RFID reader is the reader antenna size and its positioning when affixed to the handheld reader unit. It is noted that, however, the antenna design in a RFID handheld reader should fulfill several unique requirements[13]. First of all, the reader antenna in a passive RFID system should demonstrate a somewhat lower reflection coefficient level than that in a usual communication system. It is because in such a

system the backscattered signal from the tag is relatively weak, and prone to be interfered by the strong reflected signal from the reader antenna terminal. Second, in accordance with the emission regulation, the peak gain of a linear-polarized reader antenna must not exceed 6dBi in order to prevent the reader from violating the maximum allowed EIRP, i.e., 4W in North America. Moreover, regarding the public exposure to electromagnetic fields and the associated health issue, it would be beneficial if one could design a RFID handheld reader antenna with high front-toback ratio so that the absorbed electromagnetic energy by the users can be substantially reduced. We proposed the antenna having good end-fire radiation pattern and can reduced electromagnetic energy absorbed by users.

In this paper, we describe the design of compact end-fire antenna for the UHF band RFID handheld readers in North America. The antenna is composed of meandered printed dipoles, a folded reflector element and a rectangular reflector element. The proposed antenna has good radiation directional than others **RFID** antennas[11][12] and size of the proposed antenna smaller than other handheld **RFID** antenna[10][12]. The antenna configuration and design methodology will be discussed in detail in Section II. The simulated and experimental results will be illustrated in Section III. This paper is concluded with a brief summary in Section IV.

II. ANTENNA CONFIGURATION

Similar to the quasi-Yagi antenna in[14]. Figure 1 shows the geometry of proposed antenna. The final antenna parameters are optimized using the commercial electromagnetic (EM) solver HFSS 13.0, and are given in the table I. The antenna is designed for UHF RFID applications in North America, i.e., in the frequency range of 902-928MHz. The length of the driven dipole and the reflector elements optimized are for excellent simultaneously achieving input impedance matching and the dipole arms are meandered to reduce the occupied dimension. Unlike a conventional quasi-Yagi antenna, here a reflector element is in close proximity to the driven element, and is also meandered in accordance with the outline of the dipole element. Accordingly, in addition to the surface wave excited in the substrate, in the proposed design the

strong near-field coupling between the driven dipole and the reflector elements also helps improve the antenna impedance matching over a wide frequency range. Meander elements affects the resonant frequency of the antenna. The antenna elements are bent into meander shapes, suitable for fitting manufacturing form factors for a handheld RFID reader. The antenna has a high directional gain which results in the operating range around the US RFID bands (902-928MHz). Both top and bottom ground planes, which serve as reflectors in the design, keep the surface wave from propagating towards the backward direction. With such an arrangement, the backward-propagated surface wave can be substantially bounced back and further facilitates the end-fire radiation.

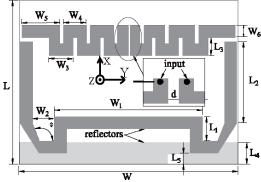


Fig. 1. The proposed antenna.

The finally chosen dimensions of the proposed antenna are illustrated in Table I.

Table I: The dimensions of the antenna (in mm)

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L	W	\mathbf{W}_1	W_2	W_3	W_4
51	80	54	8.1	9.3	9.3
W_5	L_1	L_2	L_3	L_4	φ
14.4	9	30.8	6.6	8	121°

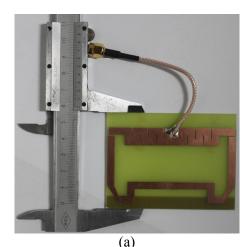
The proposed anenna is designed based on basic Yagi-Uda antenna principle, consists of two radiating elements (driver and reflector). Both elements were shaped to fit into the available dimension while maintaining their resonant frequencies in the desired band. Key parameters in the design are lengths and shapes of antenna elements and their mutual spacing. The antenna was tuned to achieve 50 ohm (RFID Reader) impedance without using any external matching circuit that will occupy additional space.

For demonstration purpose in the laboratory, the proposed antenna was designed on a 1.6 mm

FR4 substrate with a dielectric constant $\varepsilon = 4.4$ and loss tangent $\tan \delta = 0.02$. The overall dimension of the antenna is 80×59 mm, or equivalently roughly $0.24 \lambda_g \times 0.18 \lambda_g$.

III.SIMULATION AND MEASUREMENT

To verify the proposed antenna design, a prototype is fabricated as shown in Fig. 2, and the results are presented here. All the measured results are carried out in anechoic chamber using a vector network analyzer (VNA) and other microwave test instruments.



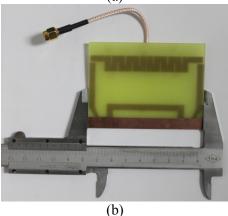


Fig. 2. Fabricated prototype of the proposed antenna. (a) top layer; (b) bottom layer.

All simulations were performed by Ansoft high-frequency structure simulation (HFSS) based on the finite-element method (FEM) [7][15]

The antenna simulated and measured magnitude of S_{11} are shown in Fig. 3. The simulation was performed by HFSS 13.0 and the measurement was taken by an Agilent performance network analyzer.

As shown in the Fig. 3, the agreement between the results in fairly good over the frequency band of interest. The simulated and measured center frequencies are given by 915 and 920 MHz, respectively. The slight frequency shift between the results can be mostly attributed to the fabrication tolerance. The measured XY-plane and XZ- plane radiation pattern and 3D radiation at 902, 915 and 928MHz are illustrated in Figs. 4, respectively. The radiation patterns are measured in a $7 \times 3 \times 3$ m³ anechoic chamber and the measurement is performed by an Agilent network analyzer along with far-field measurement software. In the measurement the connecting cables along the Bakelite support ware carefully shielded by absorbers to reduce the multireflection interference. Meanwhile, the simulated -10 dB reflection coefficient bandwidths are from 890 to 940 MHz and the corresponding measurement data are given by 900-940MHz. and the current distribution at 915MHz are shown in Fig. 5. The experimental results demonstrate that the proposed design completely complies with the stringent requirement of impedance matching imposed on a handheld reader antenna, and the operating bandwidth with reflection coefficient better than -10dB covers the whole allocated spectrum for UHF RFID applications in North America.

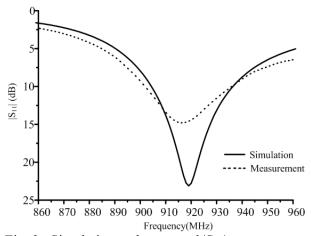


Fig. 3. Simulation and measured $|S_{11}|$.

For ease of practical applications, the studies of an important parameters of the driver meander dipoles and is also performed by simulations. One parameter is changed, while the other parameters are kept as in Table I. Figure 6 shows that the center frequency is increasing while the length of the meander dipoles varies in a range when is changed from 4.5 to 4.9mm.

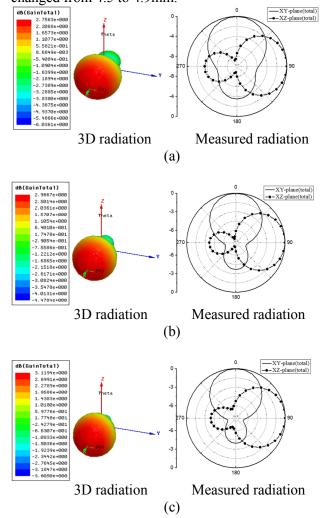


Fig. 4. 3D radiation and Measured radiation in the XY-plane and XZ-plane. (a) 902 MHz. (b) 915 MHz. (c) 928 MHz.

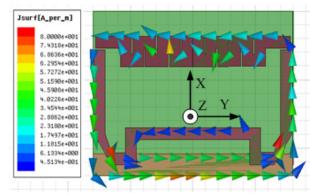


Fig. 5. The current distribution at 915MHz.

The Gain of the antenna was measured using the gain comparison method [16], where the received power of the antenna under test is compared with known gain of a standard horn antenna. The simulated and measured gain and effectively are shown in Fig. 7, variation between the simulated and measured gain is within 0.5 dB, and this may be due to higher dielectric losses of the substrate, additional loss in the surface roughness of the microstrip patch.

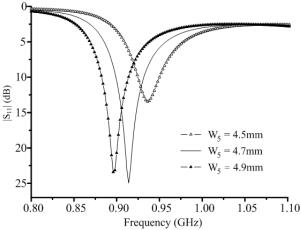


Fig. 6. Effects of varying driver mender length W₅.

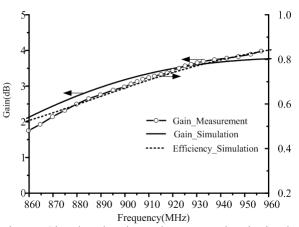


Fig. 7. Simulated gain and Measured gain in the +X direction, and Simulated efficiency of the antenna.

The Referring to the Fig. 3, measured results can be observed over the frequency band of interest. Clearly, Fig. 4 shows the radiation patterns similar to conventional Yagi radiation characteristics. The measured front-to-back ratio is at least 3 dB at 902 MHz and reaches 3.6 dB at

928MHz and remains better than 3 dB over the whole UHF RFID band from 902 to 928MHz. By adding directors elements can increase the front-to-back ratio, but these will increase the dimension of the antenna.

The measured bore sight gain is illustrated in Fig. 7. Referring to Fig. 7, the antenna gain rises steadily from 2.7dBi at 890MHz to 3.8dBi at 940MHz. The efficiency of the proposed antenna rises steadily from 62% to 80%.

IV. CONCLUSIONS

In this paper, we proposed a new printed endfire antenna for UHF RFID handheld reader applications. The new end-fire antenna is suitable for fabrication on low-cost, low dielectric constant materials such as FR-4. The new antenna is based on the conventional printed quasi-Yagi antenna, where half-wavelength dipole driver element is replaced with two meander dipoles. The input impendence of the folded dipole quasi-Yagi antenna and its resonance frequency can be tuned by properly adjusting the parameters of the mender dipoles giving freedom for optimization. The advantages of the new antenna element are that it is more compact than the conventional design and is suitable for fabrication on low-cost, low dielectric constant materials. The antenna configuration, design, simulated and measured results have been well discussed. experimental results reveal that the proposed antenna features a compact size $0.24\lambda_g \times 0.18\lambda_g$, -10 dB reflection coefficient bandwidth of 50MHz and moderate gain around 2.7 to 3.8 dB. The antenna is well designed and may find applications in a variety of circumstances and the antenna is involved in item-level UHF automation management with RFID techniques.

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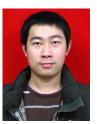
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